Analysis of MMSE-based anti-interference performance for MIMO-UWB systems

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Abstract—The combinations of MIMO (Multiple-Input-Multiple-Output) and UWB (Ultra-Wide-Band) systems greatly improve the performance of the UWB system. For the MIMO-UWB system, a new decoding scheme is proposed based on the STF (Space-Time-Frequency) coding and it can reduce the decoding complexity. The simulation results show that under the same SNR, the new method can obtain better performance of the bit error rate than the traditional ML decoding method.

Keywords-MIMO-UWB; BER; MMSE

I. INTRODUCTION

The ultra-wideband (UWB) system has many advantages, such as high transform rate, low-power, anti-multipath and anti-narrowband interference ability, and large capacity. It has been widely used in the wireless communications of high data rate and short range in recent years. In compliance with the FCC regulations, UWB is defined as nominally more than 500MHz bandwidth in the range from 3.1 to 10.6 GHz. Two standards proposed by IEEE 802.15.3a are MB-OFDM UWB and DS-UWB. In MB-OFDM scheme, the overall bandwidth of 7.5GHz is split into some smaller subbands of bandwidth 528MHz.

Multiple-input multiple-output (MIMO) technology can improve the data transfer rate, spectral efficiency, and transmission quality of the wireless communication systems, and also enhance the capacity of the system. Therefore, the combination of MIMO technology with UWB can effectively increase the data transfer rate and spectral efficiency.

There are a lot of literatures for MIMO-UWB systems [1-2] and the use of multi-antenna spatial diversity techniques greatly improve the information transfer rate. One of the key technologies of the MIMO-UWB systems is channel coding technique. In [3,4], the anti-interference ability of the system is improved through space-time-frequency (STF) coding based on $\Delta - K$ and $S - V$ channel models, but it adds the system complexity of the transmitter and receiver sides under the influence of system dimension. The pre-decoding scheme based on the dual-carrier modulation greatly simplifies the system complexity [5], but it reduces the spectral efficiency of the system. The STF coding methods based on narrowband are researched in [6,7].

In this paper, the bit error rate (BER) performance of the MIMO-UWB system is analyzed based on the S-V channel model through a simple STF coding scheme, and a low complexity MMSE algorithm for decoding is also proposed. Theoretical analysis and simulation results show that in comparison with a traditional ML decoding method, the new decoding method further reduces the BER of the MIMO MB-OFDM UWB system.

II. MIMO MB-OFDM UWB SYSTEM MODEL

A. MIMO MB-OFDM UWB system

The system with $M_t$ transmit and $M_r$ receive antennas is considered and there are $N_s$ date streams that are modulated on N subcarriers. A MIMO MB-OFDM UWB system is shown in Figure 1.

![System model for MIMO MB-OFDM UWB](image)

The OFDM symbol transmitted from the i-th transmit antenna is given by [3]

$$s_i(t) = \sum_{k=0}^{K_s-1} \Re\{x_i^s(t - kT_{sym}) \exp(j2\pi f_c^s t)\}$$  \hspace{1cm} (1)

where $K_s$ is the number of OFDM blocks for each antenna, $x_i^s(t)$ is the baseband OFDM signal to be transmitted by the transmit antenna.
the power decay factor within cluster. The average power delay is expressed as

\[ d_i^u = \begin{bmatrix} d_i(0), d_i(1), \ldots, d_i(N-1) \end{bmatrix}^T. \]

Then during the u-th OFDM block period, the complex symbol \( d_i^u(n) \) is transmitted over subcarrier \( n \) by the i-th transmit antenna, after applying OFDM modulation, which are transmitted simultaneously from all over \( M_i \) transmit antennas.

Provided that the system synchronization has been achieved in the receivers, the receive signal at the j-th receive antenna in the matrix form is given by

\[ Y_j = \sqrt{E} / M_i DH_j + Z_j \]

where \( E \) is the average energy per transmitted symbol, \( D \) is an \( M_b N \times M_b N M_i \) data matrix constructed from the STF codeword \( D_i \). \( D = [D_1, D_2, \ldots, D_{M_i}] \) and \( D_i = diag\{d_i(0), d_i(1), \ldots, d_i(M_b N - 1)\} \). Both \( Z_j \) and \( Y_j \) are Gaussian white noise signal matrix with same size. Finally, through the ML criterion, the signals \( Y_j \) in \( M_i \) receive antennas can be obtained when \( \hat{D} = \tilde{D} \) is satisfied. Here, \( \tilde{D} \) is

\[ \tilde{D} = \arg \min_{D} \sum_{j=1}^{M_i} \left\| Y_j - \sqrt{E} / M_i DH_j \right\|^2_F \]

### III. STF Coding and Decoding

#### A. STF Codes

Given that \( M_b = 2 \), \( M_1 = 2 \), \( M_2 = 1 \), and \( N = 2 \), the STF code with two OFDM blocks can be expressed as [6]

\[ D_{STF} = \begin{bmatrix} d_{1,1} & d_{1,2} \\ d_{2,1} & d_{2,2} \\ -d_{1,1}^* & -d_{1,2}^* \\ -d_{2,1}^* & -d_{2,2}^* \end{bmatrix} \]

where \( d_i = [d_{i,1}, d_{i,2}]^T = [d_{i,2}, d_{i,1}]^T \). The system channel matrix can be shown as

\[ H(h) = \begin{bmatrix} H_1(h) \end{bmatrix} \]

\[ H_1 = \begin{bmatrix} H_{1,1}^*(0) & 0 \\ 0 & H_{1,2}^*(1) \\ 0 & H_{1,3}^*(3) \end{bmatrix} \]
signal-noise ratio (SNR) 5dB relative to one with single-antenna.

\[
\begin{bmatrix}
    H_{2,1}(0) & 0 \\
    -H_{1,1}^*(1) & 0 \\
    0 & H_{2,1}(2) \\
    0 & -H_{1,1}^*(3)
\end{bmatrix}
\]

(11)

B. ML Algorithm

According to ML estimation algorithm, we can get

\[
\hat{d}_{ML} = \arg \min \| H \cdot d - y \|^2
\]

(12)

If \( H^H H = \alpha^2 \), the maximum likelihood estimation of \( d \) can be indicated as

\[
\hat{d}_{ML} = \arg \min \| H^H y - \alpha \sqrt{E/M} \cdot d \|^2
\]

\[
= \arg \min \left\{ \| H^H y - d \|^2 + (\alpha - 1)\| H \|^2 + (1 - \alpha)HH^H y \right\}
\]

\[
= \arg \min \left\{ \| H^H y - d \|^2 + (\alpha \sqrt{E/M} - 1)\| d \|^2 \right\}
\]

\[
= \arg \min \| H^H y - \hat{d} \|^2
\]

(13)

C. MMSE Algorithm

For convenient discussion, provided that \( y = H d + n \) meets, we can acquire by MMSE-BLE algorithm [8]

\[
\hat{d} = \left( H^H R_n H + R_d^{-1} \right)^{-1} H^H R_n^{-1} y
\]

(14)

where \( R_d = E\{dd^H\} \) is covariance matrix of the code vector \( d \), and \( R_n = E\{nn^H\} \) is covariance matrix of the noise vector. If \( R_d = I \) and \( n \) is white noise, equation (14) can be simplified as

\[
\hat{d} = \left( H^H H + \sigma^2 I \right)^{-1} H^H y
\]

(15)

IV. SYSTEM SIMULATION AND ANALYSIS

We can design the system simulation parameters by referring to the UWB channel model parameters that IEEE sets down (shown in Table 1), and implement the simulation of the system BER under CM1-CM4 channels respectively. In the systems with QPSK and BPSK modulation, the subcarrier number \( N=128 \), the system bandwidth 528 MHz and the guard interval \( N_g=16 \), we suppose that the carriers and symbols have strict synchronization.

Figures 2 and 3 show the system BER curves in different channels respectively. We can see from them that as the antenna number adds, the system BER performance under the both channels enhances. The performance of the MIMO-UWB system is superior to that of the SISO-UWB system obviously. For example, when the mean square error is equal to \( 10^{-4} \), the UWB system with two antennas increases

Figure 4 shows the BER performance comparison of the MIMO-UWB systems when the different decoding methods are used. It can be observed that as SNR advances, the system BER under two decoding criterions drops gradually. The system performance with the MMSE algorithm becomes better for comparison with that with the ML algorithm at the same SNR. But the decoding complexity of the MMSE algorithm becomes lower for contrast with that of the ML algorithm.
The anti-interference performance of the MIMO-UWB system is investigated in this paper. For the anti-interference performance of the UWB system, the performance comparison between the system with multiple antennas and one with single antenna is conducted. Through optimizing the STF coding and decoding ways, we introduce the MMSE algorithm which has lower complexity relative to ML algorithm. The simulation results indicate that the MMSE algorithm exceeds the ML algorithm in the anti-interference and moreover, has lower decoding complexity.

**REFERENCES**


**TABLE I. UWB CHANNEL MODEL PARAMETERS**

<table>
<thead>
<tr>
<th>Scheme</th>
<th>( \Lambda ) (1/ns)</th>
<th>( \tilde{\Lambda} ) (1/ns)</th>
<th>( \Gamma )</th>
<th>( \gamma )</th>
<th>( \sigma_\xi ) (dB)</th>
<th>( \sigma_\zeta ) (dB)</th>
<th>( \sigma_g ) (dB)</th>
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</thead>
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<tr>
<td>CM1</td>
<td>0.0233</td>
<td>2.5</td>
<td>7.1</td>
<td>4.3</td>
<td>3.3941</td>
<td>3.3941</td>
<td>3</td>
</tr>
<tr>
<td>CM2</td>
<td>0.4</td>
<td>0.5</td>
<td>5.5</td>
<td>6.7</td>
<td>3.3941</td>
<td>3.3941</td>
<td>3</td>
</tr>
<tr>
<td>CM3</td>
<td>0.0667</td>
<td>2.1</td>
<td>14</td>
<td>7.9</td>
<td>3.3941</td>
<td>3.3941</td>
<td>3</td>
</tr>
<tr>
<td>CM4</td>
<td>0.0667</td>
<td>2.1</td>
<td>24</td>
<td>12</td>
<td>3.3941</td>
<td>3.3941</td>
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</tr>
</tbody>
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Figure 4. System BER of the MMSE and ML estimation.