

Projective synchronization control of fractional-order chaotic system

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Abstract. This paper investigates projective synchronization of fractional order chaotic systems with different orders. In order to get the projective synchronization of fractional order chaotic systems with different orders, a scheme is proposed and effective controllers for synchronization are designed. Numerical simulations on examples are presented to show the effectiveness of the proposed control strategy.

Keywords: Fractional-order system; Projective synchronization; Stability theory

Introduction

In the past decades, synchronization has gained attention due to its potential applications in various fields [1-3]. Moreover, different synchronization types have been presented in chaotic systems, such as phase synchronization [4], anti-synchronization [5], projective synchronization [6] and so on. Meanwhile, effective control methods have been proposed for various synchronization types. As we known, many researchers pay their attention to the study of fractional-order chaos synchronization [7]. Fractional calculus is the generalization of the conventional calculus. It has been found that the fractional calculus has many applications in secure communication, complex network and so on. Rich dynamical behaviours exist in fractional-order system. That is to say, it is very interesting to study the projective synchronization between fractional-order systems. In this paper. We focus on demonstrate the validity of the above proposed methods for the projective synchronization on the fractional-order system.

Fractional-order definition

The fractional order is the extending concept of integer-order operator, which can be described as following:

$${}_a D_t^q = \begin{cases} \frac{d^q}{dt^q} & R(q) > 0 \\ 1 & R(q) = 0 \\ \int_a^t (d\tau)^{-q} & R(q) < 0 \end{cases} \quad (1)$$

where q is the fractional order which can be a complex number. The Riemann–Liouville fractional derivative of order $\alpha \geq 0$ of the function $f(t)$ is defined as

$$J^\alpha f(t) = \frac{1}{\Gamma(\alpha)} \int_0^t (t-\tau)^{\alpha-1} f(\tau) d\tau, \quad t > 0. \quad (2)$$

Caupto definition of the fractional derivative of the function $f(t)$ is defined as:

$$D_*^\alpha f(t) = \frac{1}{\Gamma(m-\alpha)} \int_0^t \frac{f^{(m)}(\tau)}{(t-\tau)^{\alpha-m+1}} f(\tau) d\tau, \quad t > 0. \quad (3)$$

Formula (2) will be used in this paper where α is the fractional order.

Synchronization scheme

A fractional order hyperchaotic drive system is given by

$$D^\alpha X = AX + F(X) \quad (4)$$

similarly, a response system is given as follows,

$$D^\beta Y = BY + G(Y) + U(X, Y, t) \quad (5)$$

where $X = (x_1, x_2, \dots, x_n)^T, Y = (y_1, y_2, \dots, y_n)^T$ denote state vectors, $F: R^n \rightarrow R^n, G: R^n \rightarrow R^n$ are continuous vector functions, $U(X, Y, t)$ is a controller which will be designed.

We take hyper-chaotic fractional-order Lorenz system is the drive system, it can be described as below:

$$\begin{cases} D^\alpha x_1 = a(x_2 - x_1) + x_4 \\ D^\alpha x_2 = bx_1 - x_2 - x_1x_3 \\ D^\alpha x_3 = x_1x_2 - cx_3 \\ D^\alpha x_4 = -x_2x_3 + dx_4, \end{cases} \quad (6)$$

when $\alpha = 0.96$ and $(d, h, f, r) = (10, 8/3, 28, -1)$, the fractional-order Lorenz hyper-chaotic attractor is shown in Fig.1.

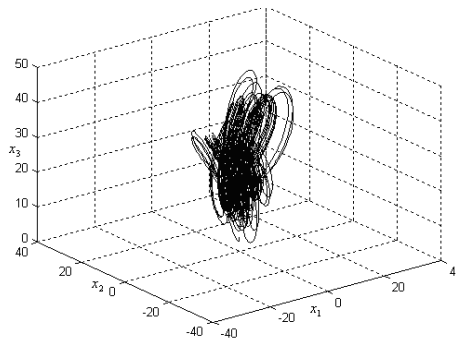


Fig.1.The attractor of Lorenz Fractional-order system with $q = 0.96$

$$\begin{cases} D^\beta y_1 = a_1(y_2 - y_1) + y_4 \\ D^\beta y_2 = d_1y_1 - y_1y_3 + cy_2 \\ D^\beta y_3 = -b_1y_3 + y_1y_2 \\ D^\beta y_4 = y_3y_2 + 0.5y_4, \end{cases} \quad (7)$$

We consider system (7) as the response system, the attractor is shown in Fig.2

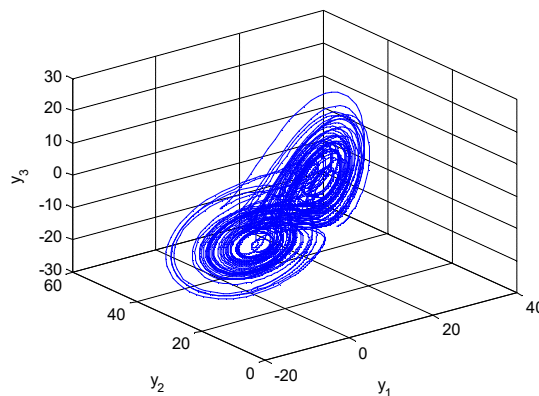


Fig.2.The attractor of Chen fractional-order hyperchaotic system with $q = 0.99$

Based on the aforementioned analysis, the systems (6) and (7) can be rewritten as following:

$$\begin{bmatrix} D^\alpha x_1 \\ D^\alpha x_2 \\ D^\alpha x_3 \\ D^\alpha x_4 \end{bmatrix} = \begin{bmatrix} -a & a & 0 & 1 \\ b & -1 & 0 & 0 \\ 0 & 0 & -c & 0 \\ 0 & 0 & 0 & d \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \\ 0 \end{bmatrix} (x_1 x_2) + \begin{bmatrix} 0 \\ -1 \\ 0 \\ 0 \end{bmatrix} (x_1 x_3) + \begin{bmatrix} 0 \\ 0 \\ 0 \\ -1 \end{bmatrix} (x_2 x_3) \quad (8)$$

$$\begin{bmatrix} D^\beta y_1 \\ D^\beta y_2 \\ D^\beta y_3 \\ D^\beta y_4 \end{bmatrix} = \begin{bmatrix} -a_1 & a_1 & 0 & 0 \\ d_1 & 0 & c_1 & 0 \\ 0 & 0 & -b_1 & 0 \\ 0 & 0 & 0 & r_1 \end{bmatrix} \begin{bmatrix} y_1 \\ y_2 \\ y_3 \\ y_4 \end{bmatrix} + \begin{bmatrix} 0 \\ -1 \\ 0 \\ 0 \end{bmatrix} (y_1 y_3) + \begin{bmatrix} 0 \\ 0 \\ 1 \\ 0 \end{bmatrix} (y_1 y_2) + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 1 \end{bmatrix} (y_3 y_2) + \begin{bmatrix} u_1 \\ u_2 \\ u_3 \\ u_4 \end{bmatrix} \quad (9)$$

According to the synchronization scheme presented in the previous discussion, we choose scaling matrix $\Lambda = \text{diag}(\lambda_1, \lambda_2, \dots, \lambda_n)$, where λ_i are the real constant. The error vector can be defined as

$$\begin{cases} e_1(t) = y_1 - \lambda_1 x_1 \\ e_2(t) = y_2 - \lambda_2 x_2 \\ e_3(t) = y_3 - \lambda_3 x_3 \\ e_4(t) = y_4 - \lambda_4 x_4 \end{cases} \quad (10)$$

The following criterion is proposed to ensure that the response system (10) effectively synchronizes the drive system (10) up to a scaling matrix.

Theorem 3.2. For a given constant scaling matrix and any initial conditions, modified projective synchronization between systems (10) and (11) will occur by the following adaptive controllers:

$$\begin{cases} u_1(t) = D^{-(\alpha-\beta)} [\lambda_1 ((a_1 - a)x_1 + (a - a_1)x_2) + x_4 - K_1 E] \\ u_2(t) = (D^{-(\alpha-\beta)} - I)(-y_1 y_3) + D^{-(\alpha-\beta)} [\lambda_2 ((d - b_1)x_1 - c_1 x_3 - x_2) + y_1 y_3 - x_1 x_3 - K_2 E] \\ u_3(t) = (D^{-(\alpha-\beta)} - I)(y_1 y_2) + D^{-(\alpha-\beta)} [\lambda_3 ((b_1 - c)x_3 - y_1 y_2) + (x_1 x_2) - K_3 E] \\ u_4(t) = (D^{-(\alpha-\beta)} - I)(y_3 y_2) + D^{-(\alpha-\beta)} [\lambda_4 (d - r_1)x_4 - y_2 y_3 - x_2 x_3 - K_4 E] \end{cases} \quad (11)$$

where $K_i (i = 1, 2, 3, 4)$ denote the i th line of matrix K . $\lambda_i (i = 1, 2, 3, 4)$ is scaling factor.

Then the error system is calculated as follows

$$D^\alpha E = (B - K)E = \begin{bmatrix} -1 & 0 & 0 & 0 \\ 0 & -1 & 0 & 0 \\ 0 & 0 & -5 & 0 \\ 0 & 0 & 0 & -0.5 \end{bmatrix} \times \begin{bmatrix} e_1 \\ e_2 \\ e_3 \\ e_4 \end{bmatrix}$$

The eigenvalues of the matrix $B - K$ satisfy $|\arg(\lambda_i)| > \frac{\alpha_i}{2} \pi = \frac{0.96\pi}{2}$. According to the stability theorems of the fractional-order system, the error system is asymptotically stable and the synchronization is guaranteed.

We choose the scaling factor as $\lambda_1 = 2, \lambda_2 = -1, \lambda_3 = 3, \lambda_4 = 1$. The synchronization errors are shown in Fig.3. In Fig.4, The phase portraits in drive and response system with different initial conditions are presented.

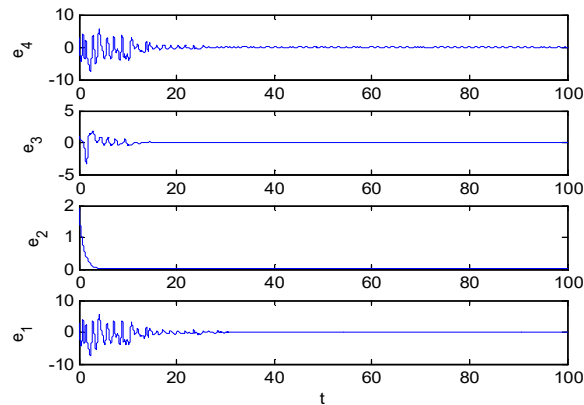


Fig.2. Synchronization errors between systems (6) and (7) with evolving time t

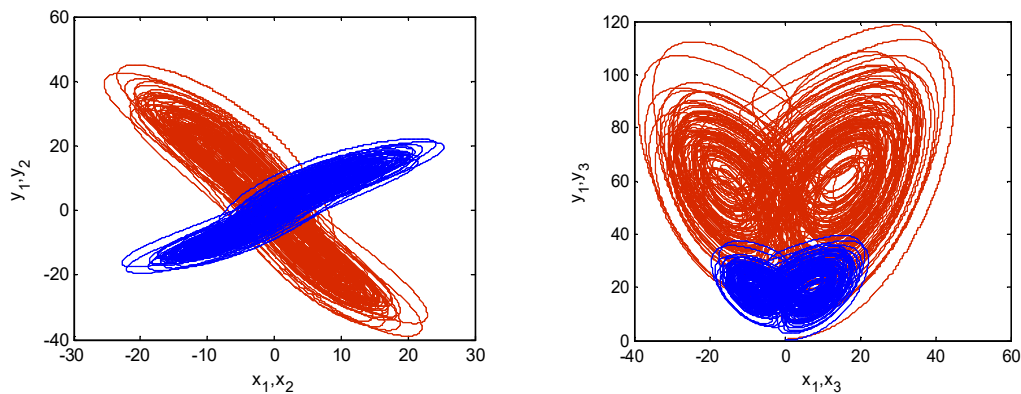


Fig.3. Phase portraits of the drive system (6) and response systems (7)

Conclusion

In this paper, projective synchronization of fractional-order chaotic system is discussed. Based on the stability theorem of fractional-order systems, projective synchronization of the fractional order chaotic system is achieved. The effectiveness of strategy is validated by numerical simulations.

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