

Synthesis of control system of recuperation of braking energy to supply grid for Metropolitan

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Abstract—The article is devoted to synthesis of an optimal regulator of the process of recovery of the braking energy of trains using an AC traction electric drive. The optimality criteria, mathematical description of the system of energy recovery, suitable for the maximum principle of L. S. Pontryagin application, were developed; the structure and parameters of an optimal controller were obtained.

Keywords—traction electric drive, recuperation energy, recuperator, the criterion of optimality, maximum principle, optimal regulator.

I. INTRODUCTION

The metro is the most promising form of mass urban transport, which fully meets the modern requirements for carrying capacity, the speed of communication, and ecology. Modern trains of the subway have traction electric drives on the basis of squirrel cage induction motors powered by an independent inverter. These trains have axial loads up to 12 tons/axle, the starting acceleration up to 1.2 m/sec^2 , the

regenerative braking throughout the speed range and require up to 70% of electricity, consumed by metro. So, the solution of the problem aimed to substantially reduce energy consumption for traction in the underground is of great importance for the economy of the region, which makes this task very urgent.

A method of saving on metro trains without the use of stored energy is described in [1]. The transfer of the braking energy of trains in the network is "bypassing" the traction rectifier through the energy recuperator from DC to AC. It is shown that the proposed method is more effective than other known conservation measures in traction drives (TED).

Fig. 1 shows a simplified single-phase diagram of energy recovery, including only the elements of TED and a traction power substation, involved in the transmission of the flow of energy from braking trains in the grid.

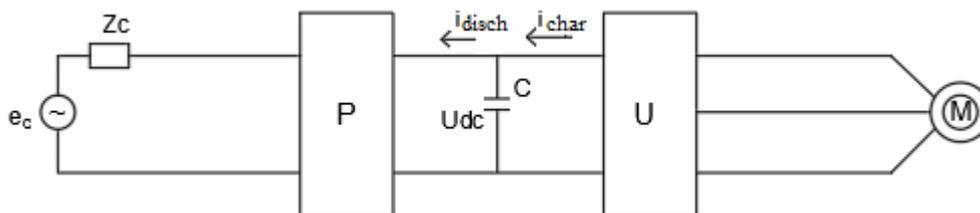


Fig.1. A simplified single-phase diagram of the electric traction drive with network elements in the regenerative braking mode.

In figure 1, there are the following notations: e_g – grid EMF; Z_g – the impedance of a grid from the AC side of recuperator P; C - capacitor in the DC link of the frequency converter with inverter U; M - the traction asynchronous motor.

The nominal voltage in the DC link is determined by the output voltage of the traction rectifier connected to AC power EMF.

Developers of the automatic control system (ACS) always have the task of selecting one optimal law, which is the most efficient in a certain sense, from the set of possible control laws. The performance indicator takes the form of quality functional $J \rightarrow \min(\max)$, which is the optimality criterion. The choice of an optimality criterion is determined by the physical meaning of the system.

Under the braking mode, the TED works as a generator, the part of the kinetic energy of the train is fed to the DC link, and capacitor C is charged by the amount:

$$\Delta U_{\text{char}}(t) = 1/R_{\text{char}} * C * \int i_{\text{char}}(t) dt. \quad (1)$$

At the same time, if $\Delta U_{\text{char}}(t)$ exceeds a certain value, recuperator P starts and the energy of the capacitor with current $i_{\text{disch}}(t)$ through P is transmitted to the network. The capacitor is discharged by the amount:

$$\Delta U_{\text{disch}}(t) = 1/R_{\text{disch}} * C * \int i_{\text{disch}}(t) dt. \quad (2)$$

In (1) and (2), $i_{\text{char}}(t)$ and $i_{\text{disch}}(t)$ are respectively the currents of charge and discharge of the capacitor; and R_{char} , R_{disch} are respectively the equivalent resistance circuits of the charging and discharging of the capacitor.

The magnitude of recovery current i_r is determined by voltage $U_d(t)$, control signal of recuperator U_y and equivalent grid parameters e_g , Z_g .

The voltage on the capacitor in the regeneration process is:

$$U_d = U_{d0} + \Delta U_{\text{char}}(t) - \Delta U_{\text{disch}}(t) = U_{d0} + \Delta U(t), \quad (3)$$

where $U_{d0} = \text{const}$ is the voltage on the DC buses in the mode of nominal thrust;

$\Delta U(t) \geq 0$ is the change in the voltage on the DC buses in the braking mode of electric trains.

It is obvious that *ceteris paribus*, the optimal transmission of the braking energy will be in the case of $\Delta U(t) = 0$, when the charge on the capacitor does not change, i.e. the current through the capacitor is 0. However, existent instability of the braking process because of a changing rate of braking commands of the driver and variability of resistance to movement due to changes in profile path in the braking process, to maintain the value of $\Delta U(t)$ close to 0, appropriate control of the recuperator $U_u(t)$ according to some rule, can be considered optimal.

To maintain a constant value of U_d , in [2], the negative feedback on this value is used; the proportional controller generates the control signal for the recuperator. As the results show, such control is far from optimal.

In the present work, the integral criterion is used to synthesize the optimal controller, i.e.:

$$J = \int \Delta U^2(t) dt \rightarrow \min. \quad (4)$$

The criterion gives the smoothing of the $\Delta U(t)$ fluctuations during braking.

II. THE MATHEMATICAL DESCRIPTION OF TEP DURING RECUPERATION

The application of existing analytical optimization methods, optimization of control systems requires mathematical description of the object of control, reflecting its main

The equation of the first order is essentially nonlinear. Its linearization to simplify the optimization problem will lead to

functions under reasonable assumptions and simplifications, which are the result of a compromise between the desire for precision reflection properties of the object and the need to simplify its mathematical description:

- taking into account only those elements of the TEP, which are involved in the process of recovery of braking energy;
- internal processes in the recuperator and the inverter are not taken into account;
- the equivalent resistance of Z_g and EMF is considered permanent;
- R_{char} and R_{disch} are permanent;
- the value of U_{d0} in the process of recovery is not changed;
- the magnitude of the charging current $i_{\text{char}} = I_{\text{char}} = \text{const}$ in interval $(0; t_T)$, defined by the mode of reference of the train;
- energy losses in the elements of the TED are not taken into account;
- grid frequency is constant;
- energy losses in the recuperator are not taken into account;
- control characteristic of the recuperator $e_R = U_u * U_d$ (at $U_d = \text{const}$) is considered to be linear;
- system variables on the AC side is represented by the rms values.
- the inertia in the DC link created by the capacitor; the inertia of the other elements of the system are not taken into account.

Taking into account the accepted assumptions, the system of equations, describing the control object, has the form:

$$\begin{aligned} d\Delta U(t)/dt &= (1/R_{\text{char}} * C) * I_{\text{disch}} - (1/R_{\text{disch}} * C) * i_{\text{disch}}(t); \\ \Delta U_{\text{char}}(t) - \Delta U_{\text{disch}}(t) &= \Delta U(t) \\ I_c(t) &= (e_c - e_r(t))/z_c; e_r(t) = U_u(t) * U_d(t); \\ \Delta U_{\text{char}}(t) &= 1/(R_{\text{char}} * C) * \int i_{\text{char}} * dt; \\ \Delta U_{\text{disch}}(t) &= 1/(R_{\text{disch}} * C) * \int i_{\text{disch}}(t) * dt; \\ d\Delta U(t)/dt &= 1/(R_{\text{char}} * C) * I_{\text{char}} - 1/(R_{\text{disch}} * C) * i_{\text{disch}}(t); \\ U_d &= U_{d0} + \Delta U. \end{aligned} \quad (5)$$

After appropriate transformations, one gets:

$$d\Delta U/dt = A - B * U_u + F * U_u^2 + D * \Delta U(t) * U_u^2, \quad (6)$$

where $A = I_{\text{char}} / R_{\text{char}} * C$; $B = e_g / R_{\text{disch}} * C * Z_g$; $F = U_{d0} / R_{\text{char}} * C * Z_g$; $D = 1 / R_{\text{disch}} * C * Z_g$.

the loss of the most important properties of the object. In addition, linearization of the multiplicative nonlinearities in

the right-hand side of the equation involves the consideration of system properties only "in the least", whereas the optimization in this case requires correction of the behavior of the object "in the large". This approach will further limit the possibility of choice of the optimization method, but allows analytical solutions.

The problem in the exact statement can be solved using calculus of variations. The maximum principle of L. S. Pontryagin is used [3-6]. The maximum principle, well suited for continuous systems, are not limited by the nonlinearities of an object or works under the integral control constraints.

On the other hand, in the general case, the maximum principle gives necessary conditions for the optimum, however, if the control process is linear and subject to additive control functions, it provides not only necessary but also sufficient conditions for optimal control.

Therefore, let us further consider the case of a linearized object.

If one assumes that the value of $U_{d0} = U_{CN} \cdot K_B$, where U_{CN} - nominal value of voltage; K_B - coefficient of transfer of an uncontrolled rectifier, when $U = \Delta U_{char} - \Delta U_{disch} > 0$, on the AC side, recuperating current occurs. The current will be transmitted by the braking energy to the grid.

With constant e_g and z_g recuperating current and, therefore, the discharge current of capacitor $i_{disch}(t)$ will be determined by value $\Delta U \cdot U_u \cdot K_R$, where K_R is a constant, depending on the design of the recuperator and grid parameters e_c and z_c , i.e:

$$i_{char}(t) = \Delta U \cdot U_u \cdot K_R.$$

In turn:

$$\Delta U = A' - \int_t (i_{disch} / R_{char} \cdot C) dt;$$

$$A' = \int_t (I_{char} / R_{disch} \cdot C) dt;$$

$$A = I_{char} / R_{disch} \cdot C.$$

Then:

$$d\Delta U/dt = A - \Delta U \cdot U_u \cdot K_R; \quad (7)$$

Linearizing the right side of equation (1), one gets:

$$d\Delta U/dt = B \cdot \Delta U + D \cdot U_u, \quad (8)$$

where $B = -U_u \cdot K_R / R_{disch} \cdot C$; $D = -\Delta U_0 \cdot K_R / R_{disch} \cdot C$; U_{u0} , ΔU_0 – coordinates of the point of linearization.

III. FORMULATION OF THE OPTIMIZATION PROBLEM

For the object, characterized by equation (8), the control action is subject to the restriction:

Then the optimal control is:

$$U_u^0 = (D \cdot x_0 / L) \cdot (e^{-\lambda_1 t} - e^{\lambda_1 (2T-t)}) / (\lambda_2 (1 + e^{\lambda_1 (2^*tT)}) + B(1 - e^{\lambda_1 (2^*tT)})) \quad (11)$$

$$\int_0^t U_u^2 = \alpha = \text{const.} \quad (9)$$

It is necessary to determine optimum control, turning into a minimum of the functional:

$$J_1 = \int_0^t (\Delta U^2 + L \cdot U_u^2) dt.$$

In the future, let:

$$x_1 = \Delta U;$$

$$\dot{x}_1 = B \cdot x_1 + D \cdot U_u;$$

$$\dot{x}_2 = x_1^2 + L \cdot U_u^2.$$

Here x_2 is the new coordinate, defined by a function:

$$x_2 = \int_0^t (x_1^2 + L \cdot U_u^2) dt.$$

Then the Pontryagin function is $p = x_2(t_T)$, when $b_1 = 0$ and $b_2 = 1$ [7,8,10].

The Hamilton function for this system is:

$$H = p_1 (B \cdot x_1 + D \cdot U_u) + p_2 (x_1^2 + L \cdot U_u^2)$$

Taking the partial derivative of H at U_u and equating it to zero, one will get:

$$p_1 \cdot D + 2 \cdot p_2 \cdot U_u = 0.$$

Taking into account condition $p_{n+1}(t) = p_2(t) = -1$ [5,9], one will get the optimal control action in the form of:

$$U_u^2 = (D/2 \cdot L) \cdot p_1.$$

Auxiliary variable p_1 is determined from the canonical equations, the general expression of which has the form:

$$\dot{p}_i(t) = -dH/dx; \quad \dot{x}_i = dH/dp_i.$$

These equations describe the conjugated system for the process control.

For the system:

$$\dot{p}_1 = B \cdot p_1 - 2 \cdot x_1 \quad (10)$$

$$\dot{x}_1 = B \cdot x_1 + D \cdot U_u^2 = B \cdot x_1 + D^2 \cdot p_1 / 2 \cdot L,$$

boundary conditions are $x_1(0) = x_0$; $p_1(t_T) = 0$.

Solving equation (10) with respect to $p_1(t)$, one will obtain:

$$p_1(t) = (2 \cdot x_0 \cdot (e^{-\lambda_1 t} - e^{\lambda_1 (2^*tT)}) / \lambda_2 (1 + e^{\lambda_1 (2^*tT)}) + B(1 - e^{\lambda_1 (2^*tT)}),$$

where $\lambda_1 = \sqrt{(B2L + D2)/L}$;

$$\lambda_2 = -\sqrt{(B2L + D2)/L}.$$

Let us note that the optimal control is changed according to the exponential law.

Lagrange multiplier L is determined by substituting (11) in limiting condition (9). This task in general is not solved; the numerical solution for a specific object is possible.

Determination of the optimal control algorithm as a function of the state variables requires the computation of $x_1(t)$ and eliminating t from the expressions for $x_1(t)$ and $U_u^o(t)$. Obviously, the general solution to this problem is extremely difficult.

However, the implementation of such system of feedback control can be performed using the above canonical equations. Measured state variable x_1 is supplied to conjugated system (10) which formulates an optimal control action.

The corresponding structural scheme obtains notation (8) and is shown in Fig. 2, where S is the Laplace operator; ΔU_u - sensor output difference $U_d - U_d^o$

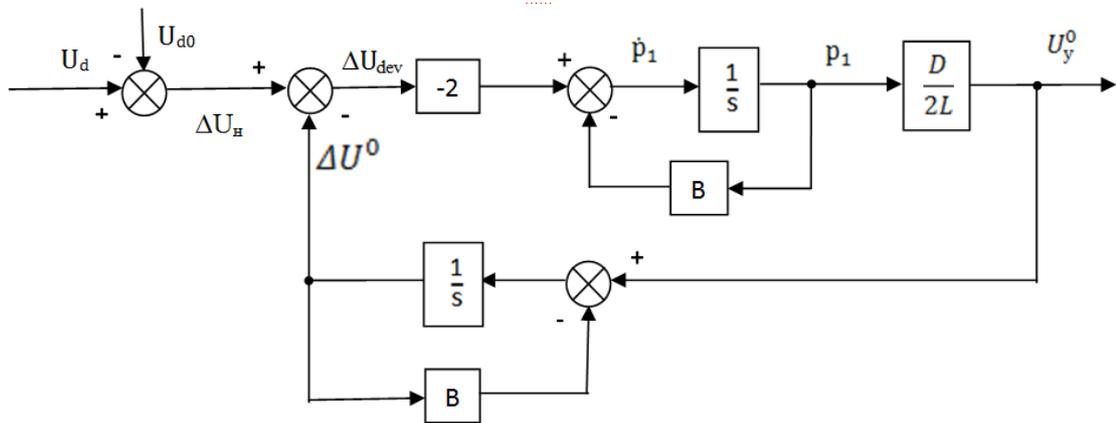


Fig. 2. Structural scheme of the optimal regulator, ΔU_{dev} - deviation of ΔU from the optimal values ΔU_0 .

The first equation of conjugated system (10) uses the signal of the measured value ΔU_N and in accordance with this equation, the upper branch of the diagram gives the value of auxiliary variable p_1 and with its help - the optimal value of control signal U_u^o .

The lower branch of the diagram corresponds to the second equation of system (10). Here on the basis of the previously obtained optimal control, an optimal value specifies impact ΔU_{ref}^o , which is compared with ΔU_u . The resulting error signal ΔU_{dev} is used in the controller to calculate U_u^o .

To check the effectiveness obtained in the present work, an optimal control modeling system recuperation of braking energy with the optimal controller is used.

The base model was applied and developed in [2]. This model embeds a model of the optimal regulator, shown in figure 3. The model of the optimal controller is represented by the lower half of figure 3. The upper half of figure 3 shows the basic elements of the recuperator (a bridge inverter with a control system "Discrete PWM Generator", a condenser and a diode element), and elements of feedback control systems of the inverter recuperator (Ref U_d , Dead Zone, Saturation, Product and others).

IV. CONCLUSION

Thus, the obtained controller has its own internal loop that helps maintain the value of ΔU at the optimum level in the process of recovery in the sense of the chosen criterion.

To check the effectiveness obtained in this work, the authors conducted a simulation of the system of traction electric drive with the use of the optimal regulator. The simulation results show that the application of the optimal regulator increases the transmission of braking energy into the power grid on average by 70-80% compared to the simple proportional control under the same conditions of the train motion (Fig.4).

Further research will be devoted to clarifying the structure and parameters of optimal controller under various conditions of the mains supply, the braking conditions of the train and other factors not considered in the present work.

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