

# High-Order Modulation Signal Generator and Its DSP Implementation

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**Abstract.** In recent years, the low-order modulation signal cannot meet current demands due to the band tension and high demand of data transfer rate. In this paper, a higher-order modulation signal generator based on software-defined radio idea was designed. The design implements an efficient plan based on DDS technology by combining the TMS320C5509 chip with CPLD, which can send and receive signals and do floating-point data operation by the method of a 16-bit fixed-point DSP calibration. The validity of the proposed method has been verified by CCS simulation of the 64 QAM, 128 QAM and 256 QAM signal constellation graph mapping.

## Introduction

Modulation signal generator is the essential part of communication and communication countermeasure, and the significant one of communication system design and electromagnetic environment simulation. It is widely used in communication, radar, measurement, and control, etc.[1]. Traditional low-order signal generator has a lot of deficiencies such as low utilization rate of spectrum, serious radiation when transmission is outside the bandwidth and poor ability of anti-interference. In the high-speed data transmission communication system, it is equally important to improve the band utilization and ensure the low symbol error rate of the system. Therefore, It is very necessary to design a DSP-based digital communication system of the high-order signal generator.

So, in this paper, the high-order modulation signal generator based on combining DSP with CPLD was proposed to meet different demands.

## Overall Structure

TMS320C5509DSP is used as the main control component of the circuit, which can modulate digital signal of input binary data, such as quadrature amplitude modulation (QAM). The EPM240T100C, connected with DSP external I/O, is used to control selection based on a 4×4 matrix keyboard. We have designed the DDS module to generate a sinusoidal sequence, whose modulus are stored into the SDRAM and DSP reads the data to choose carrier wave. DAC900E is a 10Bit, 165MSPS high-speed D/A conversion chip, and used for signal output. We use DDS as the buffer between DSP and DAC900E [2]. Besides, the design of FLASH module, loading when starts, contributes to DSP to run offline, and the asynchronous serial port module can get real-time data for transmission and display constellation and waveform on VC. The specific block diagram is shown in Fig.1.

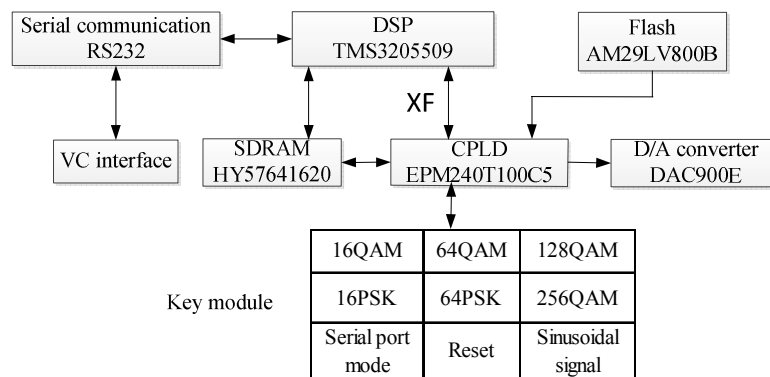


Figure 1. Finite Finite Hardware specific block diagram

## Hardware Design

**Signal Processing Module Design.** TMS320VC5509 (hereinafter referred to as VC5509) from TI company is used as DSP chip. EMP240T100C5 of Altera MAX II series as the controller and HY57641620 of 4M memory used for SDRAM as a synchronous memory chip to extend the external DSP space are implemented in this system. DSP external input clock is 10MHZ, and its working clock is 144MHZ through the internal PLL multiplier. DSP is supplied by dual power, whose standard voltage is 1.8V and I/O voltage is 3.3V. Its low 10-bit address line and 16-bit data line are connected to CPLD. External module is connected with CPLD by decoding module and the controlling signal is produced by XF of DSP I/O port which is also connected with CPLD. Bit synchronizing signal and digital baseband signals are input via the DSP external interrupt pins INT0 and GPIO0, respectively. At this time, DSP can select the corresponding digital modulation algorithm for data processing according to the key. After that, DSP places the data into the SDRAM and if XF=0, we can be informed that the CPLD has completed the digital modulation process. Finally, CPLD reads the signal that has been modulated by SDRAM and sends it to the DAC converter for external output.

**FLASH Memory Interface Technology.** Since we generally use the VC5509 of LQFP package, it has only 14-bit address line pin, and can only address the range of  $8K \times 16$  bit space. But if we want to address the AM29LV800  $512K \times 16$  bit memory, 19 address line are needed. The solution is that we can use the programmable logic device CPLD for logic decoding, that is, we control the outputs through different inputs to complete different functions. The VC5509 EMIF asynchronous interface AWE and ARE are connected with 16-bit wide FLASH OE and WE pin to control external memory for reading and writing. The low address A [0:12] of FLASH is provided by A [1:13] of VC5509 and the high address FA [13:18] is provided by CPLD. The other 16 data buses are directly connected. CPLD is connected to the low 8-bit address pins and 16-bit data pins of the VC5509 for communication and as a decoder to expand the external I / O port. The byte select pin and the reset pin of the FLASH device are connected to CPLD and controlled by CPLD.

The reading and writing process of the FLASH is given as follows. DSP clock and CPU initialization is the first step, and then we program the FLASH into the working state and set the EMIF global control register and CE1 related control register. According to the function of AM29LV800B, we can erase the entire chip memory. In the bootloader mode, VC5509 writes data to the address of 0x400000 to 0x800000 and the 0xFFFF of Flash must be stored the first address of the boot table. Afterwards, under the control of CCS3.3, the Boot table is written to FLASH. Finally, the contents written into FLASH are read out to check in which VC5509 would automatically implement when it has been powered.

## Software Design

**DDS implementation.** A direct digital frequency synthesizer consists of phase accumulator, adder, waveform ROM, D/A converter and low pass filter (LPF) structure. The reference oscillator provides a stable time base for the system and determines the frequency accuracy of the DDS. It provides the clock to the NCO which produces a discrete-time and quantized version of the desired waveform at its output (often a sinusoid) whose period is controlled by the digital word contained in the Frequency Control Register. The digital waveform is converted to an analog waveform by the DAC. The output reconstruction filter rejects the spectral replicas produced by the zero-order hold inherent in the analog conversion process[3][4].

In the paper, the determination of the frequency, amplitude, carrier source whose phase is adjustable, are the foundation and key of the design of the high-order modulation signal generator. In VC5509 system, each time the implementation of a timer interrupt program would output a data point. The timer interval is 50 DSP clocks, so the DDS system clock is 2MHz. The low 30 bits of accumulator in designed program are used as accumulator registers. As the DDS system clock  $f_s = 2\text{MHz}$ , according to the DDS principle, we can determine some data such as the system frequency resolution  $\Delta f = 2\text{MHz}/2^{30} = 0.00186\text{Hz}$ , output frequency  $f = m \times 0.00186\text{Hz}$ , theoretical output maximum frequency is 1MHz and actual maximum output frequency is 40% of the system clock, that is 800KHz. To complete the waveform reconstruction, DDS algorithm sample the sine discrete table fixed in the FLASH under certain frequency. In order to improve the speed, we usually place the initialization of the discrete sine table to the internal high-speed RAM to run.

**Quadrature Amplitude Modulation (QAM) Theory.** Quadrature amplitude modulation (QAM) is a kind of modulation method with high spectrum utilization. Compared with other modulation methods, it has a higher data transmission rate, and it is very competitive in the future communication field. In order to adapt to the changing wireless channel, high-order modulation signal can greatly improve the efficiency of data transmission.

QAM is a vector modulation. First, the input bit (usually the Gray code) maps to a complex plane (constellation), in which the bit can be transformed into complex modulation symbols. Then the I and Q component of the symbol (corresponding to the real and imaginary part, that is the horizontal and vertical one of the complex plane) are taken with amplitude modulation, which, respectively, correspond to the two carrier ( $\cos \omega_c t$  and  $\sin \omega_c t$  generated by the DDS) [5]. Therefore, it will exponentially increase the spectrum utilization compared to amplitude modulation (AM). QAM is a Joint technique of magnitude and phase, which utilizes the amplitude and phase of the carrier to transmit the information bits, so that higher band utilization can be achieved with specific and identical conditions[5]. Ideally, if carrier states of the M-QAM is M, it can modulate  $\log_2 M$  bits. The signal form of M-QAM is given as

$$s_m(t) = \text{Re} \left[ (A_{mc} + jA_{ms}) g(t) e^{j2\pi f_c t} \right] = A_{mc} g(t) \cos(2\pi f_c t) - A_{ms} g(t) \sin(2\pi f_c t) \quad (m=1, 2, \dots, M, 0 \leq t \leq T) \quad (1)$$

where  $A_{mc}$  and  $A_{ms}$  are the quadrature carrier carrying information,  $g(t)$  is the baseband waveform pulse after forming. QAM signal modulation principle is shown in Fig.2.

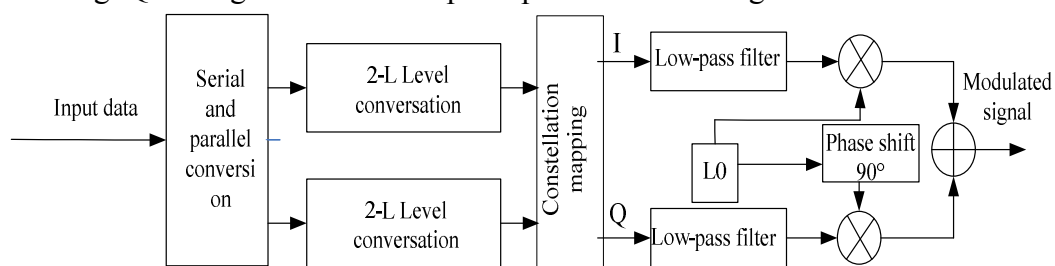


Figure 2. Finite QAM modulation principle diagram

The input data is divided into two paths by serial-to-parallel conversion, and then become  $A_{mc}$  and  $A_{ms}$  by 2-level to L-level ( $\pm L = \pm \sqrt{\log_2 M}$  level) transformations, which determine the position of the constellation point of the modulated QAM signal in the signal constellation.

**VC Interface Design.** In order to fully embody the "software radio", high-order modulation signal generator is connected with the PC through asynchronous serial port, and MFC, working in the Windows system, is used for the design of interface. We can select the required modulation signal and change the amplitude, phase, and frequency of the carrier by the design.

MFC (Microsoft Foundation Classes), a class library provided by Microsoft, encapsulates the Windows API in the form of a C++ class, like a large number of Windows handle wrappers and many Windows built-in controls and components of the wrapper class[6], and includes an application framework to reduce the workload of application developer.

DSP will send modulated data to the PC through the interruption of the asynchronous serial port. Function signalgenerate () serves as the signal generation in Visual C++, and setpoint (x) represents the function of waveform drawing with the MFC class library. At the same time, DSP can read the carrier frequency, amplitude and phase set in the interface according to the button.

### DSP Simulation Result

In this experiment, 16QAM, 64QAM, 128QAM and 256QAM quadrature amplitude modulation signals are simulated in CCS3.3. The sampling frequency  $F_s = 32000\text{Hz}$ , carrier frequency  $F_c = 1000\text{Hz}$ , symbol rate  $f_{16QAM} = 9600\text{ bit/s}$ ,  $f_{64QAM} = 38400\text{ bit/s}$ ,  $f_{128QAM} = 76800\text{ bit/s}$ , and  $f_{256QAM} = 153600\text{ bit/s}$ . The test results are shown below. Constellation points of all three quadrant are evenly distributed and amplitude, phase are correct as well. Constellations of MQAMs are shown as Fig.3.

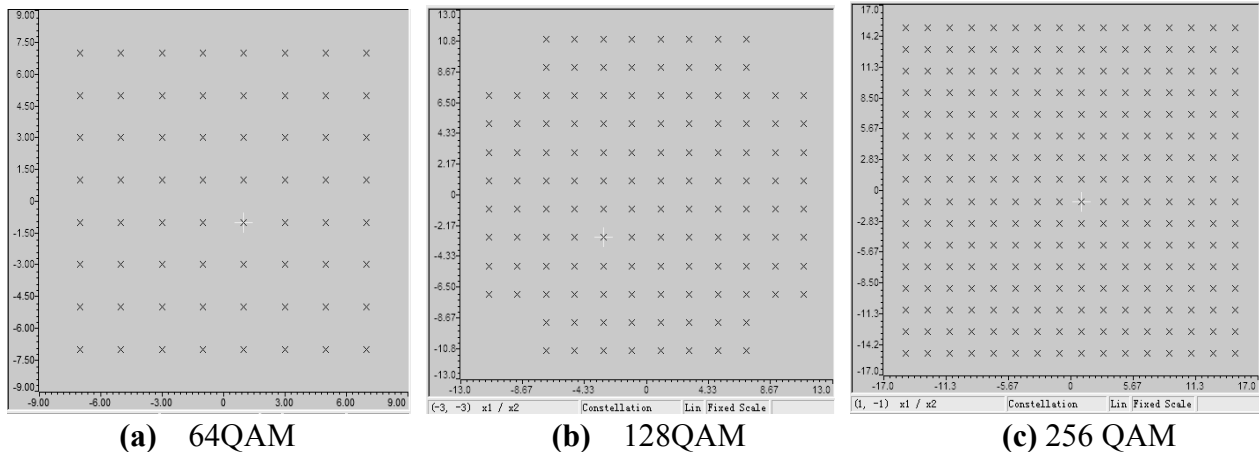


Figure 3. Finite Constellations of MQAMs

### Conclusion

Based on "software radio", this paper presents and implements a design of a high-order modulation signal generator designed by DSP, which can obtain the required modulation signal by key selection, realize the interface operation of human-computer interaction, and achieve offline operation using FLASH self-loading method. In addition, the MQAM signals are simulated on the CCS, and the results are correct and compelling, which verifies that the high-order modulation signal generator is practical and achievable.

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