Parameters Turning of ADRC based on Neural Network

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Abstract-1. Objective: Though ADRC controller shows strong robust character and adaptability, it still exist a lot of shortages, such as the frequency characteristic, stability... etc. Those were not solved in the theories. And the most difficulty of ADRC's application in the industry is that the parameters are too more and the parameters' adjusting is difficulty. The engineers go short of the experience of the parameters' adjusting for the controller. To solve the problems that the active disturbance rejection controller has too many parameters and it is very difficult to calculate a set of optimal parameters without determinate turning algorithms. 2. Methods: Neural network can get the gradient information of the controlled object, then use gradient descent method to modify nonlinear combined parameters online, so that the ADRC controller has the ability of self-learning, enhanced adaptive ability of ADRC. 3. Results: The simulation results show that, the ADRC controller, which has good dynamic and static features, improves the design efficiency. The feasibility and effectiveness of this method is further verified.

Key words-tracking differentiator; gradient information; ADRC controller; parameters turning; self-learning

I. INTRODUCTION

ADRC was firstly proposed by Han in [1–2], advantages and disadvantages of classical PID are investigated from its basic principles. A new type controller-Active Disturbances Rejection Controller (ADRC) with excellent characters is constructed via some links with especial functions, such as Tracking Differentiator (TD), Extended States Observer (ESO) and Nonlinear PID (NPID), which are based on the nonlinear control mechanisms. Furthermore, the new active disturbances rejection control technique is formed. The new controller possesses the following characteristics: the algorithm is simple; the adjustment of the parameters is easier. ADRC is a new non-linear algorithm used in different fields in recent years, it has been proposed and developed for almost two decades, and its applications can be found in lots of literature in recent years. It is a control method that does not depend on the accurate mathematical model of the unknown object. By real-time estimation and compensations of the internal and external disturbances of system, combining with nonlinear control strategy, it can get better static and dynamic performances, strong robustness and adaptability. Since ADRC does not depend on the accurate model of the system, it is very robust against parameter variations, disturbances and noises, not only in some operation areas but also in the whole working area. Many novel design methods [3-10] for the auto-disturbance-rejection controllers (ADRC) were proposed to deal with the difficulty in the parameter regulation of the ADRC. Bring forward ADRC controller based on NN. The kind of method can adaptively adjust
the parameters of nonlinear error feedback control rule NLSEE.

II. ADRC STRATEGY

Classical proportional-integral-derivative (PID) is a particular primitive and simplified implementation of the basic principle in error-based feedback control, which focuses on eliminating the control error by using the current, past and future states of the feedback error. There is a question, that if the load changes in a very large range, we can not change the parameters online to meet the system requirements. However, the ADRC method which doesn’t depend on system model can estimate system requirements. ADRC not only has the same advantages of fast response and strong robustness as traditional PID control theory, but also gives a new control theory and control method which is widely applied for its excellent system performances.

The ADRC consists of three parts, a nonlinear TD which is used to obtain the ideal transient process of the system, an ESO which estimates all the disturbances by the system output, and then the ADRC compensates the disturbance according to estimated values, a nonlinear state error feedback (NLSEF) which is used to get the control input of the system. The structure of ADRC controller is shown in Figure.1.

Figure 1. The structure of ADRC system

A. TD

One feasible second-order TD can be designed as

\[\begin{align*}
    x_1(k+1) &= x_1(k) + h_1 \cdot e_1(k) \\
    x_2(k+1) &= x_2(k) + f(\alpha, \delta, x_1(k), x_2(k), r, h_1)
\end{align*}\]  
(1)

Where x1 denotes the control objective, r is the speed factor which decides tracking speed. The greater value of r is, the faster transition process will be. Here h0 is the filtering factor which makes an effort of filtering. As we know, decreasing the integration step will make a great effect on limiting the noise. When the integration step is fixed, increasing the filtering factor will make the filtering effect better. The function is defined as follows:

\[f(\alpha, \delta, x_1(k), x_2(k), r, h_1) = \begin{cases} 
    -r \left( \frac{d}{\alpha} \right) & |\alpha| \leq d \\
    -r \text{sign}(\alpha) & |\alpha| > d
\end{cases}\]  
(2)

\[\begin{align*}
    d &= r \cdot h_0 \\
    d_0 &= dh_0 \\
    y(k) &= x_1(k) - y_0 + h_0 \cdot x_2(k) \\
    \alpha_0 &= \sqrt{d^2 + 8r} |y(k)| \\
    x_2 + \frac{y(k)}{h_0} |y(k)| &\leq d_0 \\
    \alpha &= \begin{cases} 
        \text{sign}(y(k)) \cdot (\alpha_0 - d) & |y(k)| > d_0 \\
        \frac{y(k)}{2} & |y(k)| \leq d_0
\end{cases}
\]

The TD is such a nonlinear component which provides transition process for expected input v and differential trajectory of x1 and its differential x2. TD has the ability to track the given input reference signal with fast response and no overshoot.

B. ESO

\(f(x_1, x_2, \omega(t), t)\) generally includes three parts: modeling dynamics, uncertain dynamics (or uncertain accelerations) and disturbance; it is difficult to get the exact model of \(f(x_1, x_2, \omega(t), t)\) or its approximation. ESO is used to estimate \(f(x_1, x_2, \omega(t), t)\) in real-time and to make adjustments at each sampling point in a digital controller. Here \(f(x_1, x_2, \omega(t), t)\) is considered an extended state for the system, \(x_3\) is the uncertain \(f(x_1, x_2, \omega(t), t)\), and its differential \(\alpha(t)\).

We can use the following nonlinear observer to estimate \(x\) and \(f(x_1, x_2, \omega(t), t)\):

\[\begin{align*}
    e_1(k+1) &= e_1(k) + \beta_1 \cdot f_{\alpha}(e_1(k), \alpha_1, \delta) + \beta_2 \cdot f_{\delta}(e_1(k), \alpha_1, \delta) + \beta_3 \cdot f_{\alpha}(e_1(k), \alpha_1, \delta) + \beta_4 \cdot f_{\delta}(e_1(k), \alpha_1, \delta)
\end{align*}\]  
(3)

\[\begin{align*}
    e_2(k+1) &= e_2(k) + \beta_1 \cdot f_{\alpha}(e_2(k), \alpha_2, \delta) + \beta_2 \cdot f_{\delta}(e_2(k), \alpha_2, \delta) + \beta_3 \cdot f_{\alpha}(e_2(k), \alpha_2, \delta) + \beta_4 \cdot f_{\delta}(e_2(k), \alpha_2, \delta)
\end{align*}\]  
(4)

\[\begin{align*}
    e_3(k+1) &= e_3(k) + \beta_1 \cdot f_{\alpha}(e_3(k), \alpha_3, \delta) + \beta_2 \cdot f_{\delta}(e_3(k), \alpha_3, \delta)
\end{align*}\]  
(5)

\[\begin{align*}
    z_1(k+1) &= z_1(k) + h(z_1(k) - \beta_0 e_1(k)) \\
    z_2(k+1) &= z_2(k) + h(z_2(k) - \beta_0 f_{\alpha}(e_1(k), \alpha_1, \delta) + \beta_0 f_{\delta}(e_1(k), \alpha_1, \delta)) \\
    z_3(k+1) &= z_3(k) - h \beta_0 f_{\alpha}(e_1(k), \alpha_1, \delta)
\end{align*}\]  
(6)

C. Nonlinear combination

State error feedback control law generates control signal \(u\) for system by using the error between the output of ESO and TD. The errors are combined in nonlinear manners; large errors is corresponding to lower gains, and small errors is corresponding to higher gains.

where \(e_1, e_2\) are the output errors. PID, as a control law, employs a linear combination of present, accumulative, and predictive forms of the tracking error and has, for along time, ignored other possible combinations that are potentially much more effective. As an alternative, we propose the following nonlinear functions:

\[\begin{align*}
    e_1(k+1) &= x_1(k+1) - z_1(k+1) \\
    e_2(k+1) &= x_2(k+1) - z_2(k+1) \\
    u(k+1) &= \beta_1 \cdot f_{\alpha}(e_1(k+1), \alpha_1, \delta) + \beta_2 \cdot f_{\delta}(e_1(k+1), \alpha_1, \delta) + \beta_3 \cdot f_{\alpha}(e_1(k+1), \alpha_1, \delta) + \beta_4 \cdot f_{\delta}(e_1(k+1), \alpha_1, \delta)
\end{align*}\]  
(7)

ADRC parameter Self-learning algorithm adopts optimization index:
\[ J = \frac{1}{2} (r(k) - y(k))^2 \]  
\[ \text{error} = r(k) - y(k) \]  

\[ \beta_1, \beta_2 \text{ can use gradient descent method to modify nonlinear combined parameters.} \]  

\[ \beta_i(k) = \beta_i(k-1) - \eta \frac{\partial J}{\partial \beta_i} \]  
\[ \beta_i(k) = \beta_i(k-1) - \eta \frac{\partial J}{\partial \beta_i} \]  

Using the neural network [3] or TD to recognize Jacobian information \( \frac{\partial y}{\partial u} \).

III. THE EXPERIMENTAL RESULTS

Simulation example use transfer function of reference[4]:  
\[ G(s) = \frac{88.46}{4.11s^2 + 2.63s + 1} \]

ADRC parameters:  
step=0.1ms; TD: \( r = r_i = 35 \); \( h_i = h = 0.1 \text{ms}; \) ESO: \( \alpha_1 = 0.5; \) \( \alpha_2 = 0.25; \) \( \delta = 0.1 \text{ms}; \) \( \beta_0 = 600; \) \( \beta_0 = 120000; \) \( \beta_0 = 8000000; \) \( b_0 = 41.9; \) NLSEF: \( \alpha_3 = 1; \) \( \alpha_4 = 0.5; \) \( \beta_1 = 70; \) \( \beta_3 = 1.0; \)

IV. CONCLUSIONS

Theoretical analysis and simulation results show that the proposed neural network application to extract the gradient information of the object being controlled in ADRC parameter online modification is effective and feasible, improve the quality control system, reduce the design difficulty.

REFERENCES